

Current Flow Controlling Hybrid DC Circuit Breaker

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Abstract—This paper proposes a new device by combining features of interline dual H-bridge current flow controller with the core idea of hybrid HVDC circuit breaker for meshed HVDC grid application. The proposed device can substitute 2 dc circuit breakers at a dc bus with at least 2 adjacent transmission lines. In addition to the current interruption action, the current in one of the adjacent lines can be controlled by the embedded current flow controller. The system level behavior of the proposed current flow controlling hybrid dc circuit breaker is similar to that of typical hybrid dc circuit breaker and the interline dual H-bridge current flow controller. The operation principles of the proposed device are introduced and analyzed in this work. The components ratings are compared to the existing solution and the functionality of proposed device is verified by simulation.

Index Terms—DC Circuit Breaker, Meshed dc Grid, Fault Protection, Current Flow Controller.

I. INTRODUCTION

DC power transmission technology was revived since the middle of twenties century by realization of the first dc cable link between mainland Sweden and Gotland island in 1953 [1], [2]. Recent advances in the converters technology and the need for transmitting bulk amount of electrical energy over long distances have brought about the HVDC technology as a cost-efficient and reliable solution [2]. Nowadays, multi-terminal HVDC (MT-HVDC) grid concept is widely considered by both academia and industry due to the increasing demand for collecting the offshore wind energy [1]. While MT-HVDC grid offers several advantages, its operation faces a few drawbacks. The MT-HVDC grid protection against dc short circuit fault has been identified as the major difficulty due to inability of most of the VSC topologies in blocking the dc fault current [2]–[4]. Even for VSC technologies able to block the fault current (employing full bridge cells), there is a need to use additional protective devices when complex MT-HVDC grids with more than one protection zones are considered [1]. Moreover, the interruption of dc current is technically difficult due to the lack of natural zero crossing [5], [6]. In addition to the protection problem, a meshed HVDC (M-HVDC) grid as a complex form of the MT-HVDC grid may face power flow control problems. Typically, the power flow in the M-HVDC grid is controlled by regulating the converters' dc side voltage considering the transmission line impedance. Due to the M-HVDC grid topology, there are multiple paths for the current to circulate between two different nodes. Consequently, some of the transmission lines can be overloaded because of their lower impedances [7].

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Over the last decade, the dc current interruption problem has been addressed by the introduction of several dc circuit breaker (DCCB) topologies [5], [8]–[11]. Among the proposed devices, the hybrid dc circuit breaker (HCB) merges the fast turn-off feature of semiconductor switches with the low-loss performance of metallic contacts and hence is technically highly attractive [12], [13]. The HCB requires hundreds of semiconductor switches in its main breaker (MB) branch to tolerate the system voltage [14], [15], hence its implementation cost is expected to be high. The number of required semiconductor switches for protection of a dc bus with two adjacent lines would be comparable to that of a modular multilevel converter (MMC) station [14].

Furthermore, several current flow controller (CFC) devices have been introduced to solve the power flow problem in M-HVDC grid [16]–[21]. The series CFCs demonstrate less power losses and reduced cost due to their lower voltage requirement and hence the fewer number of switches. The interline dual H-bridge CFC with reduced number of switches can be considered as one of the most efficient CFCs [18].

The coordinated operation of HCBs and CFCs is expectable in the future M-HVDC grids. As an alternative solution, this paper proposes a new device, which benefits from the core idea of typical HCB [12] and possesses an embedded CFC [18] and therefore can be named as current flow controlling circuit breaker (CFCCB). The CFCCB has three ports and can connect a dc bus to two adjacent transmission lines. The CFCCB can regulate the current in one of the adjacent lines and upon receiving a trip command can interrupt the fault current and isolate the faulty line from the dc bus. The proposed approach requires fewer number of semiconductor switches as compared to the typical approach. Moreover, the ratings of required surge arresters in DCCB part can be reduced remarkably by employing the proposed device.

This paper is organized as follows: the CFCCB topology and operation principles are detailed in section II. The case study model is explained in section III and section IV provides the analysis of CFCCB operation. The simulation results are presented and discussed in section V. A comparison between the typical approach and the proposed CFCCB is included in section VI and the paper is concluded in section VII.

A. CFCCB topology

The topology of proposed CFCCB is shown in Fig. 1. The CFCCB has three terminals and can connect a dc bus to two adjacent transmission lines.

II. CURRENT FLOW CONTROLLING DC CIRCUIT BREAKER

The CFCCB consists of main breaker (MB_i) units, ultra fast disconnectors (UFD_i), surge arresters (SA_i), disconnectors (DS_i), seven semiconductor valves and one capacitor. Moreover

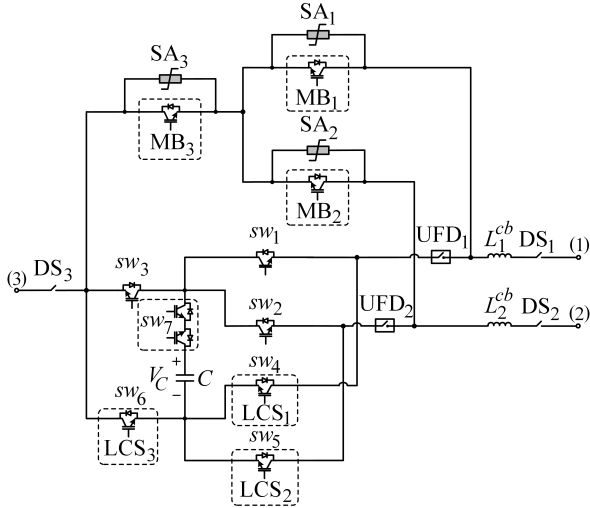


Fig. 1. Current flow controlling dc circuit breaker topology

(L_1^{cb}) and (L_2^{cb}) are employed as current limiting inductors. The MB units consist of several semiconductor switches with antiparallel diodes in series connection. The semiconductor switches are connected in one direction and hence the MB units are unidirectional. UFD₁-UFD₃ are assumed to be similar to the UFDs in typical HCB [12]. DS₁-DS₃ are standard high voltage disconnectors to provide the electrical isolation after CFCCB action. As shown in Fig. 1, sw₁-sw₆ together with capacitor C form an interline dual H-bridge CFC with reduced number of switches [18]. However, sw₄-sw₆ are also exploited as load commutation switches (LCSs). Therefore, sw₄-sw₆ are also referred as LCS₁-LCS₃ in this paper.

A. Operation principles

1) *CFC bypassed mode*: In this mode, the CFCCB does not control the current and only maintain the power flow between the dc bus and the adjacent lines. The equivalent representation of CFCCB in the CFC bypassed mode is depicted in Fig. 2(a). Note that the semiconductor switches are represented by ideal switches in Fig. 2. In this mode, sw₁-sw₃ and sw₇ are opened whereas sw₄-sw₆ are closed. DS₁-DS₃ and UFD₁-UFD₂ are closed while MB₁-MB₃ are opened. It can be seen in the figure, the current can flow between the terminals of CFCCB irrespective of its direction.

2) *Active CFC mode*: The equivalent representation of CFCCB in the active CFC mode is depicted in Fig. 2(b). In this mode the CFCCB controls the current in one of the adjacent lines by operating its embedded dual H-bridge CFC. Depending on the current direction, the desired voltage can be generated by selecting a suitable set of states of switches. Table I illustrates the switching states for both negative and positive currents. The capacitor voltage is represented by V in Table I. The current can be controlled using a PI and second order compensator. The linearized average model of the CFC represented by a couple of voltage sources can be used to design the current control system [7]. The embedded CFC operation modes and control scheme have been extensively investigated in [18] and [7]. As shown in Fig. 2, i_1 , i_2 and i_3 are the currents flowing through terminal 1, 2 and 3 of CFCCB, respectively. Based on the switching states in Table I, i_2 can be controlled only by applying PWM signal to sw₆ assuming the following scenario:

- i_3 is incoming current into the CFCCB and i_2 and i_1 are outgoing currents.
- The currents are positive.

TABLE I
SWITCHING STATES FOR POSITIVE AND NEGATIVE CURRENT SCENARIOS [7]

Positive Currents					
Set	sw ₁	sw ₆	sw ₂	V ₃₁	V ₃₂
1	0	0	0	-V	-V
2	0	0	1	-V	0
3	0	1	0	0	-V
4	0	1	1	0	0
5	1	0	0	0	0
6	1	0	1	0	+V
7	1	1	0	+V	0
8	1	1	1	+V	+V
Negative Currents					
Set	sw ₃	sw ₄	sw ₅	V ₃₁	V ₃₂
9	0	0	0	+V	+V
10	0	0	1	+V	0
11	0	1	0	0	+V
12	0	1	1	0	0
13	1	0	0	0	0
14	1	0	1	0	-V
15	1	1	0	-V	0
16	1	1	1	-V	-V

3) *Fault mode*: The CFCCB can receive three independent trip commands including two line faults and one dc bus fault trip commands. Upon receiving a trip command the corresponding terminal(s) of CFCCB must interrupt(s) its(their) current(s). Note that the CFCCB may enter to the fault mode either when it operates in the CFC bypassed or in the active CFC modes.

a) *Fault on adjacent transmission line*: Assume a permanent short circuit fault happens on the line connected to the terminal 1 of CFCCB. Hence the terminal 1 of CFCCB should trip and interrupt the fault current. It is assumed that the fault incept at time t_f and the trip command is received by the CFCCB at time t_1 . At time t_1 , sw₁-sw₃ should be opened and consequently sw₄-sw₆ must be closed at time t_2 . This action redirects the fault current path into the LCS units and prevents the capacitor from charging or discharging by reducing its current to zero. Thereafter sw₇ can be opened at time t_3 in zero current to ensure that the capacitor is disconnected from the system. Upon opening of sw₇, MB₁-MB₃ must be closed at time t_4 . The equivalent representation of this stage is shown in Fig. 2(c). As can be seen in the figure, the fault current is shared between MB and LCS units. At time t_5 , sw₄-sw₆ should be opened and commutate the current into the MB units. At this stage, the current in UFD units is almost zero. Therefore, UFD₁ can be opened at time t_6 in order to isolate the terminal 1 of CFCCB (Faulty line corresponding terminal) from the dc bus and the other adjacent line. Consequently, sw₄-sw₆ should be closed at time t_7 . Fig. 2(d) shows the equivalent representation of CFCCB at time t_7 . Note that, the current cannot flow through LCS₁ due to the open state of UFD₁. Finally, MB₁-MB₃ opens and redirect the currents into the SA₁-SA₃ at t_{br} . The time period between t_7 and t_{br} is named as current sharing stage. The current in SA₁ will be zero due to the conduction of antiparallel diodes of MB₁. Fig. 3 illustrates the sequential operation of CFCCB in the line fault mode. The operation sequence for a fault on the line connected to terminal 2 of CFCCB is similar to the explained case. However, in the latter scenario UFD₂ must be opened instead of UFD₁. To provide the electrical isolation,

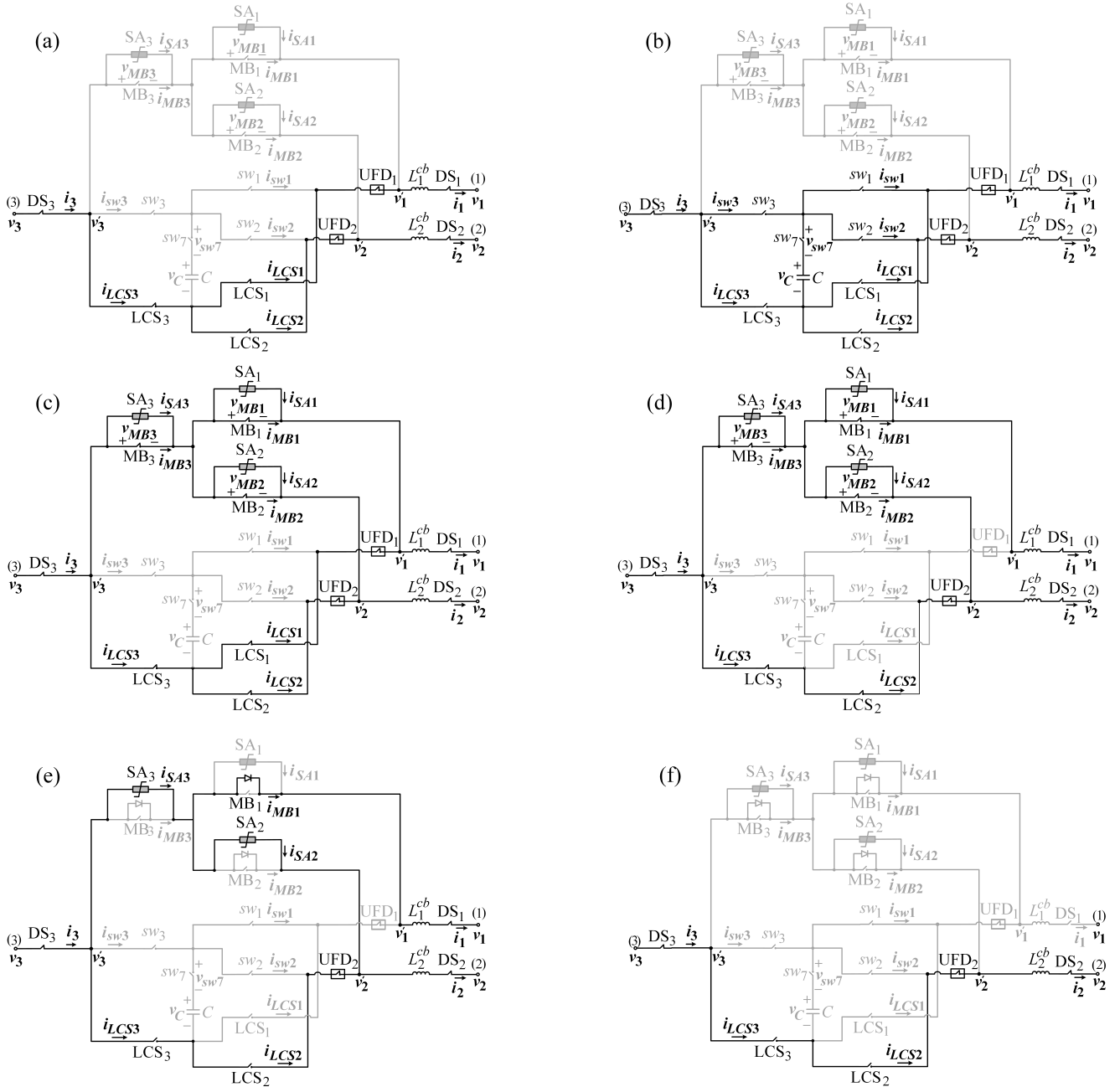


Fig. 2. Operation stages of current flow controlling hybrid dc circuit breaker

the corresponding terminal disconnector (DS) can be opened.

b) Fault at dc bus: Upon detection of a permanent dc bus fault, all adjacent lines must be isolated from the dc bus. The operation sequence for a bus fault interruption is illustrated in Fig. 3. As it is shown in the figure, the operation sequences for line and bus faults are the same until time t_5 . To interrupt a dc bus fault, it is necessary to open both UFD₁ and UFD₂ at time t_6 . For the bus fault interruption there is no current sharing stage and MB₁-MB₃ can be directly opened at time t_{br} to interrupt the fault current and redirect it into the surge arresters. Note that the current in SA₃ will be zero due the conduction of the antiparallel diodes of MB₃. The bus fault current interruption is expected to be faster than the line

fault scenario due to the lack of current sharing stage. Finally, the electrical isolation can be provided by opening DS₁-DS₃.

c) Recloser mode: The recloser mode might be required before completely opening of the CFCCB. The CFCCB can be reclosed by reclosing MB₁-MB₃ after opening the faulty line corresponding UFD. The equivalent circuits of reclosing mode are equal to Fig. 2(d) and (f). Finally, in case of a non-permanent fault, the faulty line corresponding UFD can be again closed and the CFCCB shifts to its normal conduction state by closing sw₄-sw₆ and opening MB₁-MB₃.

III. CASE STUDY MODEL

The current interruption operation of CFCCB does not depend on the grid topology. Therefore, an average model

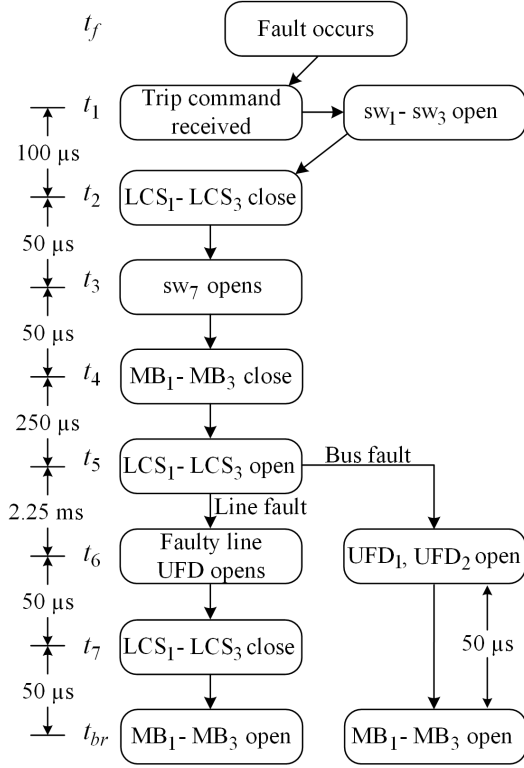


Fig. 3. Line and bus faults isolation process

of a three-terminal meshed grid is selected for performing the analysis and simulation. The three-terminal grid model is shown in Fig. 4. The transmission lines are modeled using lumped T-equivalent model. The parameters of test system are illustrated in Table V. The analysis and simulations are carried out for both proposed and typical schemes. As illustrated in Fig. 4, in the typical scheme, a CFC is installed at bus B_1 to regulate the current in line 12. In addition, two HCBs are installed between two adjacent lines (12 and 13) and the CFC. In this study, the HCBs from [12] are considered. A detailed schematic of the HCB is illustrated in Fig. 5. In the proposed scheme, the CFC and the HCBs are substituted by the CFCCB (see Fig. 4). The CFC control system is designed based on [7] for both typical and proposed schemes. The parameters of CFC are the same for both mentioned schemes and are illustrated in Table V.

IV. ANALYSIS

The operation of embedded CFC has been analyzed in [7], [18]. Therefore, only the current interruption mode of CFCCB is considered in this section. In order to clarify the differences between the proposed and the typical methods the analysis is simplified by considering the following aspects:

- The permanent dc fault and prompt fault interruption strategy are considered [15], [22].
- Voltage at dc buses are assumed to be constant during the DCCB operation time [23].
- Transmission line and dc bus short circuit faults are modeled by a voltage source, whose value is equal to the system steady-state voltage value in normal condition and it changes to 0 V as soon as a fault happens.

Considering the aforementioned assumptions, the grid model can be simplified by eliminating the connection between buses B_2 and B_3 . The simplification can be done due to

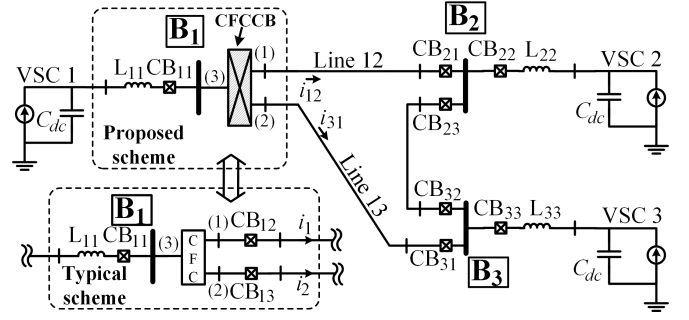


Fig. 4. Average model of three-terminal grid

assumed constant dc bus voltage during the fault condition. The simplified model for analysis is depicted in Fig. 5.

A. Impact of the embedded CFC on the fault current

The embedded CFC is composed of six switches with their antiparallel diodes and one bidirectional switch. Due to arrangement of the switches and the capacitor, the CFC cannot block the fault current until the capacitor is charged up to the nominal line voltage. The voltage rating of the capacitor lies in the range of few kilo volts. Therefore, the CFC capacitor must be disconnected from the fault current path to prevent it from being charged or discharged. The capacitor can be retained by opening sw_1 - sw_3 while closing sw_4 - sw_6 and then opening sw_7 . Considering the embedded CFC structure, it can be found out that all the switching states may only change the fault current path inside the CFC and the fault current may charge or discharge the capacitor. Hence, during the fault clearing time period (which lies in range of few milliseconds) the CFC has almost no impact on the fault current. When $0 < t \leq t_2$, the current flows through sw_1 - sw_7 depending on their state. Therefore considering the scope of paper the analysis considers the current equations for $t_2 < t \leq t_{br}$.

B. Transmission line fault F_1 and CFCCB

A low impedance ($R_{fault} \approx 0 \Omega$) pole-to-ground fault occurs on line 12 at point F_1 at $t=0$ s. The voltage at fault location (v_{F1}) becomes zero after fault occurs. The following equations can be given considering the initial conditions and assumptions:

$$\begin{aligned} v_{F1}(0) &= V_{dc}, \\ v_{F1}(0^+) &= 0, \\ v_j(t) &= V_{dc}; \quad 0 < t \leq t_{br}, \quad j=1,2,3, \\ i_j(0) &= I_{pre,j}; \quad j=1,2,3. \end{aligned} \quad (1)$$

In (1) $v_{F1}(t)$, $v_j(t)$ and $I_{pre,j}$ represent the voltage at fault location, dc bus B_j voltage and pre-fault current of port j of the CFCCB. t_{br} represents the current interruption instant. The current at port 3 can be given as follows:

$$i_3(t) = \begin{cases} i_{LCS3}(t), & t_2 < t \leq t_5 \\ i_{MB3}(t), & t_5 < t \leq t_7 \\ i_{MB3}(t) + i_{LCS3}(t), & t_7 < t \leq t_{br} \\ i_{SA3}(t) + i_{LCS3}(t), & t_{br} < t \leq t_e \end{cases}, \quad (2)$$

and for port 1:

$$i_1(t) = \begin{cases} i_{LCS1}(t), & t_2 < t \leq t_5 \\ i_{MB1}(t), & t_5 < t \leq t_7 \\ i_{MB1}(t), & t_7 < t \leq t_{br} \\ i_{MB1}(t), & t_{br} < t \leq t_e \end{cases}, \quad (3)$$

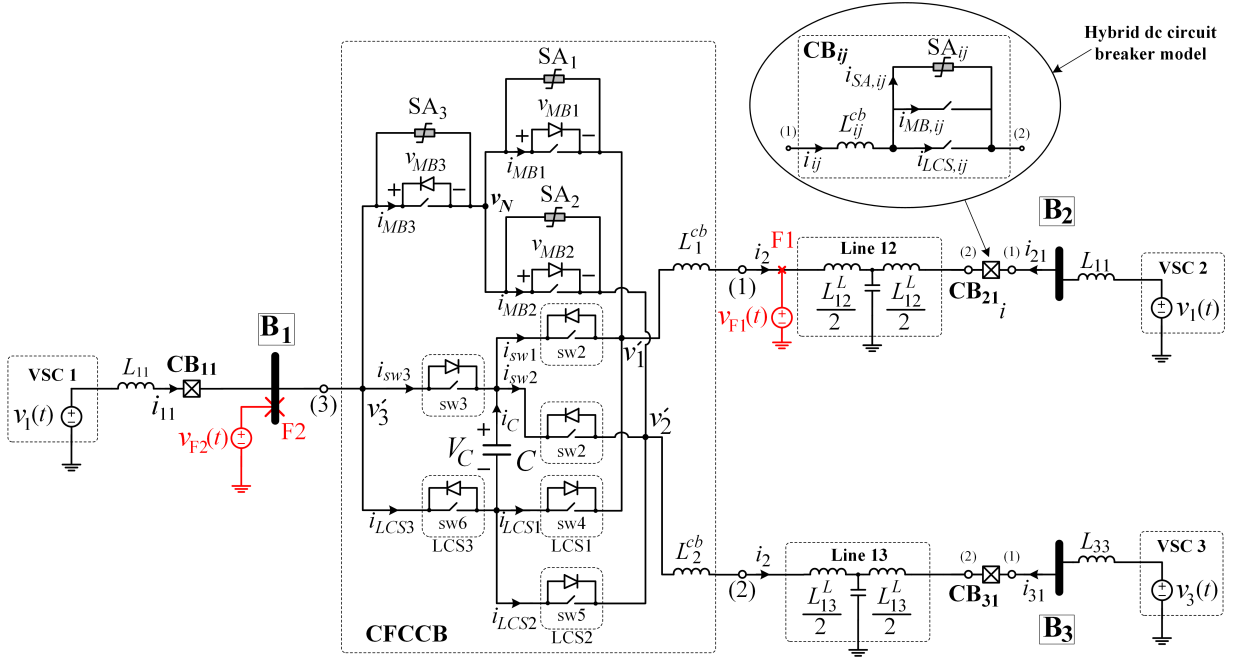


Fig. 5. Equivalent circuit of test system in presence of CFCCB

and for port 2:

$$i_2(t) = \begin{cases} i_{LCS2}(t), & t_2 < t \leq t_5 \\ i_{MB2}(t), & t_5 < t \leq t_7 \\ i_{MB2}(t) + i_{LCS2}(t), & t_7 < t \leq t_{br} \\ i_{SA2}(t) + i_{LCS2}(t), & t_{br} < t \leq t_e \end{cases} \quad (4)$$

We replace the sum of half of line $1j$ inductance (L_{1j}^L) and corresponding port current limiting inductor value (L_{j-1}^{cb}) by L'_{1j} :

$$L'_{1j} = L_{j-1}^{cb} + \frac{1}{2}L_{1j}^L \quad \text{for } j=1,2 \quad (5)$$

$$L'_{11} = L_{11} + L_{11}^{cb}$$

Therefore, considering the impact of transmission line lumped T-equivalent model capacitance during the short fault clearing time period, the initial rate of rise of current at port 1, which is equal to that of fault current can be given by:

$$\frac{di_1(0^+)}{dt} = \frac{V_{dc}}{\left((L'_{11} \parallel L'_{13}) + L_{11}^{cb} + L_{12}^f \right)} \quad (6)$$

where L_{12}^f represents the inductance between the CFCCB and the fault location. The current derivative at the other ports of CFCCB can be given as:

$$\left| \frac{di_2(0^+)}{dt} \right| = \frac{L'_{11}}{(L'_{11} + L'_{13})} \left| \frac{di_1(0^+)}{dt} \right|, \quad (7)$$

$$\left| \frac{di_3(0^+)}{dt} \right| = \frac{L'_{13}}{(L'_{11} + L'_{13})} \left| \frac{di_1(0^+)}{dt} \right|.$$

The current in MB and LCS units can be obtained using (2)-(4), (6) and (7).

1) *Main breaker (MB) units*: When $t_4 < t < t_{br}$ MB₁-MB₃ are closed. Therefore, assuming instantaneous current commutation at $t=t_5$ the current in MB₁ can be given as:

$$i_{MB1}(t) = I_{pre,1} + \left| \frac{di_1(0^+)}{dt} \right| t \quad (8)$$

The maximum current in MB₁ (I_{max}^{MB1}) happens at $t=t_{br}$. The current in MB₂ and MB₃ when $t_4 < t < t_7$ can be given by:

$$i_{MBi}(t) = I_{pre,i} + \text{sgn}(i-2.5) \left| \frac{di_i(0^+)}{dt} \right| t; \quad i=2,3. \quad (9)$$

The maximum current in MB₂ and MB₃ is reached at $t=t_7$. When $t_7 < t < t_{br}$, the current is shared between two mentioned MBs and their current will be decreased.

2) *Load commutation switch (LCS) units*: As was explained, the LCS switches conduct the current in two periods of time: i) $t_2 < t < t_5$ ii) $t_7 < t < t_{br}$. In the first stage (when $t_2 < t < t_5$), the current in LCS_j for $j=1$ and for $j=2,3$ holds the same equations as (8) and (9), respectively. The maximum current in the first stage in the LCS units happens at $t=t_5$. The second maximum of current in the LCS units occurs at time t_{br} . Fig. 2(d) shows the equivalent circuit of the CFCCB when $t_5 < t \leq t_{br}$. The MBs have several IGBTs in series whereas the LCSs have only few IGBTs. The on-state voltage drop on an MB can be hundred times larger than the on-state voltage drop of an LCS. Hence, the voltage drop on the LCSs can be neglected against that of MB units. Therefore, the following equation can be given considering Fig. 2(d):

$$v'_2 \approx v'_3 \quad (10)$$

$v'_1-v'_3$ are illustrated in Fig. 2. Based on (10), MB₂, MB₃ can be considered as parallel branches during the mentioned time period and their currents will be almost equal. The absolute value of current in MB_j at $t=t_7^+$ can be given by:

$$|i_{MBj}(t_7^+)| = \frac{|i_{MB3}(t_7^-)| + |i_{MB2}(t_7^-)|}{2}; \quad j=2,3 \quad (11)$$

Using (2), the current in LCS_n can be given as follows:

$$i_{LCS3}(t_7^+) = i_{LCS1}(t_7^+) = \frac{|i_{MB3}(t_7^-)| - |i_{MB2}(t_7^-)|}{2} \quad (12)$$

The time $t_{br}-t_7$ lies in the range of tens of microseconds. However, using the obtained current derivative in (6), the second maximum of current in LCS_j for $j=2,3$ can be obtained as:

$$I_{max2}^{LCSj} = i_{LCSj}(t_7^+) + \frac{di_1(0^+)}{dt} \cdot \frac{t_{br}-t_7}{2} \quad (13)$$

3) *Surge arresters (SA)*: Surge arresters have a non-linear voltage-current characteristic. Only for comparison purposes, SA's parameters are approximated by assuming its voltage to

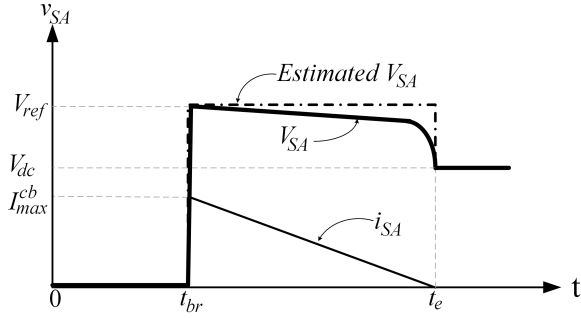


Fig. 6. Surge arrester approximated current and voltage

be constant until its current falls to zero for both proposed and typical schemes. Fig. 6 illustrates the voltage and current approximation method used for the SAs. It is assumed that the SA current reaches its maximum instantaneously and then decreases linearly. This method is used to identify the maximum possible energy absorption in the surge arresters. Neglecting the practical mismatch between $V-I$ characteristics of surge arresters, the current can be given as:

$$\begin{aligned} |i_{SAj}| &= \left| \frac{i_{MB1}}{2} \right|; \quad j=2,3 \\ |i_{SA1}| &= 0, \end{aligned} \quad (14)$$

The current in SA_1 is zero due to the conduction of antiparallel diodes in MB_1 . Considering (10) it can be assumed that SA_2 and SA_3 operate in parallel connection. The rated voltage of each surge arrester is assumed to be equal to V_r . The transient interruption voltage (TIV) across MB_2 and MB_3 can be given by:

$$TIV = V_{dc} + \left((L'_{11} \parallel L'_{13}) + L_1^{cb} + L_{12}^f \right) \frac{I_{max}^{MB1}}{t_e - t_{br}} \quad (15)$$

The current in SA reaches zero when its voltage falls below its rated voltage. The maximum required time for SA current to fall to zero ($t_e - t_{br}$) can be given as:

$$(t_e - t_{br}) \leq \left((L'_{11} \parallel L'_{13}) + L_1^{cb} + L_{12}^f \right) \frac{I_{max}^{MB1}}{V_r - V_{dc}}, \quad (16)$$

The maximum absorbed energy in all the surge arresters holds:

$$E_{SA,T} = \int_{t_{br}}^{t_e} V_r \cdot i_1(t) dt \quad (17)$$

Consequently, the maximum absorbed energy in SA_j can be given as:

$$\begin{aligned} E_{SAj} &= \frac{V_r I_{max}^{MB1} (t_e - t_{br})}{4}; \quad j=2,3, \\ E_{SA1} &= 0. \end{aligned} \quad (18)$$

C. DC bus fault F_2 and CFCCB

As shown in Fig. 5, a low impedance pole-to-ground fault ($R_{fault} \approx 0 \Omega$) occurs at dc bus B_1 at time 0 s. The initial conditions and study assumptions are similar to (1) and also similar approach to subsection IV-B is used for analysis. The current at port j for various time periods can be given by:

$$i_j(t) = \begin{cases} i_{LCSj}(t), & t_2 < t \leq t_5 \\ i_{MBj}(t), & t_5 < t \leq t_{br} \\ i_{SAj}(t), & t_{br} < t \leq t_e \end{cases}; \quad j=1,2,3 \quad (19)$$

The current derivative at ports of CFCCB can be given by:

$$\begin{aligned} \frac{di_3(0^+)}{dt} &= V_{dc} \frac{L'_{12} + L'_{13}}{L'_{12} L'_{13}}, \\ \left| \frac{di_i(0^+)}{dt} \right| &= \frac{V_{dc}}{L'_{1(i+1)}}; \quad i=1,2. \end{aligned} \quad (20)$$

1) *Main breaker (MB) units*: The currents in MB units when $t_5 < t \leq t_{br}$ can be given as:

$$i_{MBi}(t) = \begin{cases} I_{pre,i} - \frac{V_{dc}}{L'_{1(i+1)}} t; & i=1,2 \\ I_{pre,i} - V_{dc} \frac{L'_{12} + L'_{13}}{L'_{12} L'_{13}} t; & i=3 \end{cases} \quad (21)$$

The maximum current in MB_j (I_{max}^{MBj}) is reached at $t = t_{br}$.

2) *Load commutation switch (LCS) units*: Despite the line fault scenario, the current in LCS units has one maximum at $t = t_5$ and can be given as:

$$I_{max}^{LCSi}(t) = \begin{cases} I_{pre,i} - \frac{V_{dc}}{L'_{1(i+1)}} t_5; & i=1,2 \\ I_{pre,i} - V_{dc} \frac{L'_{12} + L'_{13}}{L'_{12} L'_{13}} t_5; & i=3 \end{cases} \quad (22)$$

3) *Surge arresters (SA)*: The SA current can be given as:

$$\begin{aligned} |i_{SAj}| &= \left| \frac{i_{MB3}}{2} \right|, \quad j=1,2 \\ |i_{SA3}| &= 0. \end{aligned} \quad (23)$$

Depending on the length of adjacent lines, the absorbed energy in the surge arresters of the CFCCB and also the energy absorption time ($t_{ej} - t_{br}$) can be different for each surge arrester. The range of energy absorption time can be given as:

$$t_{ej} - t_{br} \leq \frac{L'_{1(j+1)} I_{max}^{MBj}}{V_r - V_{dc}}, \quad j=1,2 \quad (24)$$

where t_{ej} is the time when the current in SA_j becomes zero. Due to conduction of antiparallel diode D_3 , the current in SA_3 remains zero and consequently the absorbed energy in SA_3 is also zero. The absorbed energy in SA_j can be given by:

$$E_{SAj} = \frac{V_r I_{max}^{MBj} (t_{ej} - t_{br})}{2}, \quad j=1,2 \quad (25)$$

D. Typical scheme

In the typical scheme, after receiving a trip command by the corresponding HCB at time t_1 , its MB unit is closed. Thereafter, the LCS unit opens at time t_2 and consequently MB unit opens at time t_3 [24]. Similar line and bus fault scenarios to subsection IV-B to subsections IV-B and IV-C are considered. For sake of brevity, only the most relevant equations are included in this section.

1) *Transmission line fault and HCB*:

a) *Load commutation switch (LCS) units*: The LCS current in CB_{12} reaches its maximum at $t = t_2$ whereas the current in LCS unit of CB_{13} reaches its maximum at $t = t_{br}$. The maximum current in LCS unit of CB_{12} can be given by:

$$I_{max}^{LCS12}(t) = I_{pre,12} + \left| \frac{di_{12}(0^+)}{dt} \right| t_2 \quad (26)$$

and for LCS unit of CB_{13} :

$$I_{max}^{LCS13}(t) = I_{pre,13} - \left| \frac{di_{13}(0^+)}{dt} \right| t_{br} \quad (27)$$

b) *Main breaker (MB) units*: It is assumed that only the MB unit of corresponding HCB of the faulty line is activated. Therefore, the current in MB units of other HCBs remain zero. The current in MB unit of CB_{12} for $t_2 < t \leq t_{br}$ can be given as:

$$I_{max}^{MB12}(t) = I_{pre,12} + \left| \frac{di_{12}(0^+)}{dt} \right| t \quad (28)$$

The maximum current in the MB unit in CB_{12} (I_{max}^{MB12}) is reached at time t_{br} .

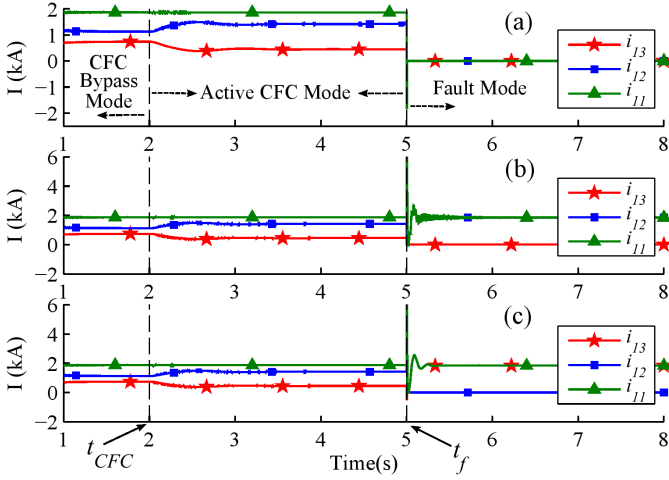


Fig. 7. Transmission lines and dc bus currents during fault at: a) dc bus B_1 , b) line 13, c) line 12

c) *Surge arresters (SA)*: The currents in surge arresters of all the HCBs are zero except the faulty line HCB (SA_{12}). The absorbed energy in the surge arrester can be given by:

$$E_{SA12} = \frac{V_r I_{max}^{MB12} (t_e - t_{br})}{2} \quad (29)$$

2) DC bus fault and HCB:

a) *Load commutation switch (LCS) units*: During the bus fault, all the adjacent HCBs of the faulty dc bus are activated. The maximum current in the LCS unit of all adjacent HCBs (I_{max}^{LCS1i}) can be given as:

$$I_{max}^{LCS1i} = I_{pre,1i} - \frac{V_{dc}}{L'_{1i}} t_2; \quad i=2,3 \quad (30)$$

b) *Main breaker (MB) units*: The currents of MB units for $t_2 < t \leq t_{br}$ can be given as:

$$i_{MB1i}(t) = I_{pre,1i} - \frac{V_{dc}}{L'_{1i}} t; \quad i=2,3 \quad (31)$$

The maximum current in the MB unit (I_{max}^{MB1i}) is reached at $t = t_{br}$.

c) *Surge arresters (SA)*: The absorbed energy in SA_{1j} can be given by (32).

$$E_{SA1j} = V_r I_{max}^{MB1j} (t_{ej} - t_{br}), \quad j=2,3 \quad (32)$$

V. SIMULATION RESULTS

In this section the simulation results of the three-terminal grid (Fig. 4) for line 12 and dc bus fault scenarios are presented and compared to the obtained numerical values from the analysis of simplified grid model in section IV. The simulations are carried out using PSCAD. The non-linear $V - I$ characteristic of surge arresters are also considered. The transmission lines are protected by standard overcurrent protection scheme. The line fault trip command is sent to the CFCCB or the corresponding HCB when the line current exceeds 2.8 kA. The dc buses are protected by the differential protection scheme. In this scheme, when the sum of incoming and outgoing currents at a dc bus becomes non-zero, the dc bus trip signal is activated.

1) *Power flow*: The currents flowing from the dc bus and the transmission lines in presence of the CFCCB are depicted in Fig. 7. Fig. 7(a), (b) and (c) depict the currents for dc bus B_1 , line 13 and 12 fault scenarios, respectively. The CFCCB operates in CFC bypassed mode for $0 < t < 2$ s. Thereafter the CFCCB changes its operation mode to active CFC mode at time $t_{CFC} = 2$ s. In all scenarios, the fault happens at time

$t_f = 5$ s. The behavior of CFCCB has been found out to be similar to the typical scheme during normal operation and fault condition from the grid point of view.

2) *Transmission line Fault*: To consider the most sever scenario, a low impedance pole-to-ground fault (100 m Ω) is placed very close to the CFCCB (distance from CFCCB is equal to 0 km.) on line 12 at $t = 0$ s. In the typical approach CB_{12} and CB_{21} and in case of the proposed CFCCB, CB_{21} and port 1 of the CFCCB should trip. Fig. 8 and Fig. 9 show different waveforms for the proposed and typical schemes, respectively. The important numerical values obtained from simulation and analysis are also illustrated in Table II. The interrupted current in the CFCCB is almost 5% larger than the interrupted current in the typical scheme due to the additional time that is considered in the modeling of current sharing stage in the CFCCB. As shown in Fig. 8(b) and Fig. 9(b), the current in sw_1 - sw_3 in both schemes are almost equal. However, a large difference in the current of sw_4 - sw_6 can be observed in Fig. 8(c) and Fig. 9(c). The absolute value of current in sw_4 and sw_6 in the typical scheme reaches almost 10 kA whereas in sw_5 and sw_6 in the proposed scheme does not exceed 5 kA. Fig. 8(d) depicts the CFCCB capacitor voltage and current and also the voltage across sw_7 . The absolute value of voltage across sw_7 does not exceed 2.5 kV. As shown in Fig. 8(f) and Fig. 9(f), the fault current is redirected into two surge arresters (SA_2 and SA_3) in CFCCB whereas it is handled by one surge arrester in the typical method. Since the rated voltage of surge arresters in CFCCB are equal to that of HCBs, the maximum voltage across MB units in both methods are equal (Fig. 8(g) and Fig. 9(g)). Fig. 8(h) and Fig. 9(h) show the absorbed energy in the surge arresters in both methods. The amount of absorbed energy in each surge arrester in the CFCCB reaches almost 7.2 MJ whereas it reaches approximately 14 MJ in the typical scheme. The results confirm that the absorbed energy in the surge arrester in the typical scheme is almost equal to twice the absorbed energy in each surge arrester in the proposed scheme.

3) *DC bus Fault*: A low impedance pole-to-ground fault (100 m Ω) is placed at bus B_1 . In the typical scheme CB_{11} , CB_{12} and CB_{13} and in the proposed scheme CB_{33} and all the ports of CFCCB should trip. Fig. 10 and 11 depict the results for the proposed and the typical schemes, respectively. The most relevant numerical values obtained from analysis and simulation are illustrated in Table III. The obtained approximated values from analysis are close to the values obtained from simulation. Fig. 10(a) and 11(a) show that the behavior of both schemes from system point of view are similar. As can be seen in Fig. 10(b) and 11(b) the current in sw_1 - sw_3 for both schemes are equal. Fig. 10(c) and 11(c) show that the current in sw_4 - sw_6 in the typical scheme may reach higher values as compared to the proposed scheme. The maximum current in MB_{12} and MB_1 and also in MB_{13} and MB_2 are equal (see Fig. 10(e) and 11(e)). The absolute value of current in MB_3 of CFCCB reaches almost 1.8 kA, which is higher than the current in MB units of CB_{12} and CB_{13} . However, this does not necessarily mean that the antiparallel diodes of MB_3 should be rated for higher current than the antiparallel diodes of MB_1 and MB_2 . In fact, MB_1 and MB_2 may be required to carry higher currents during a line fault and should be rated for that. Fig. 10(f) and 11(f) illustrate that the maximum current in the surge arresters of both schemes are equal. In contrary with MB_1 and MB_2 , the current in MB_3 does not redirected to the surge

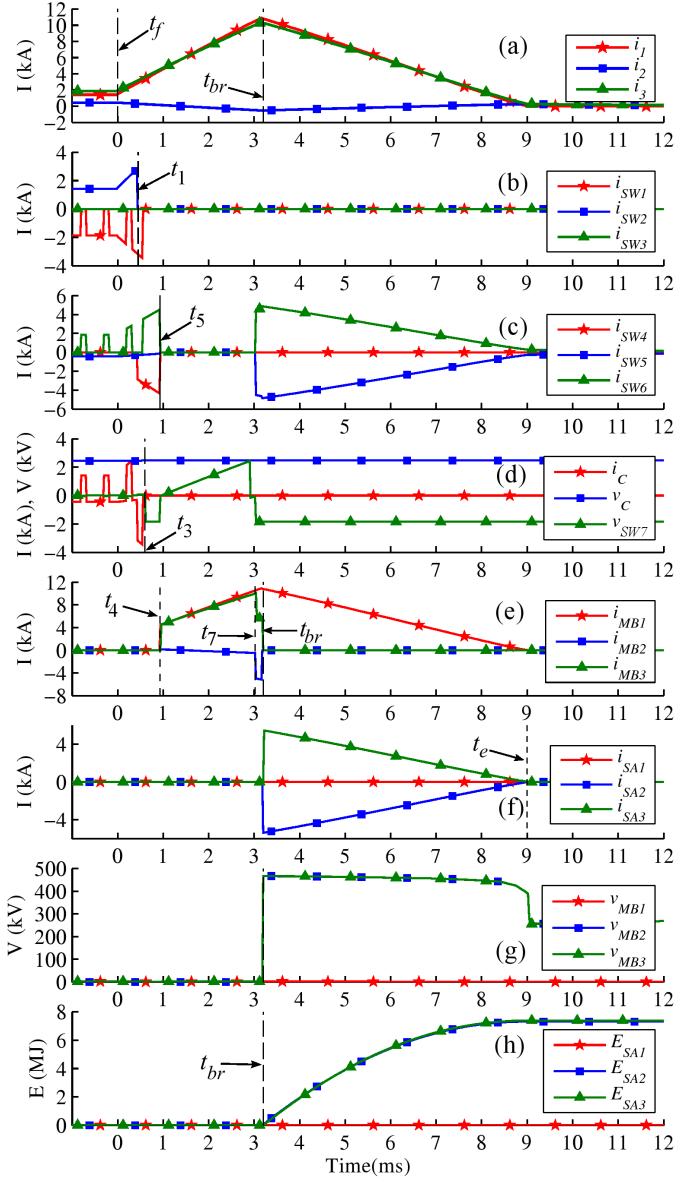


Fig. 8. CFCCB waveforms for fault on line 12

TABLE II
CFCCB AND HCB PARAMETERS DURING LINE FAULT

Parameters CFCCB (HCB)	CFCCB		HCB	
	Analysis	Simulation	Analysis	Simulation
$I_{LCS1} (I_{LCS12})$ [kA]	4.18	4.28	3.20	3.28
I_{MB1}^{max} [kA]	4.48	4.62	-	-
$I_{MB1}^{max} (I_{MB12})$ [kA]	11.10	10.9	10.43	10.33
$E_{SA1} (E_{SA12})$ [MJ]	0	0	16.98	13.69
$E_{SA2} (E_{SA13})$ [MJ]	8.99	7.31	0	0
E_{SA3} [MJ]	8.99	7.38	-	-

arrester due to explained reason in section IV. The absorbed energy in the surge arresters in the CFCCB and the typical scheme are depicted in Fig. 10(h) and 11(h), respectively.

VI. COMPARISON

The proposed scheme is compared to the typical scheme in this section. As shown in Fig. 4, to fully protect a dc bus with 2 adjacent lines and one converter station with an asymmetric monopole HVDC configuration, 3 HCBs are required in the typical scheme. The number of HCBs can be doubled in symmetric

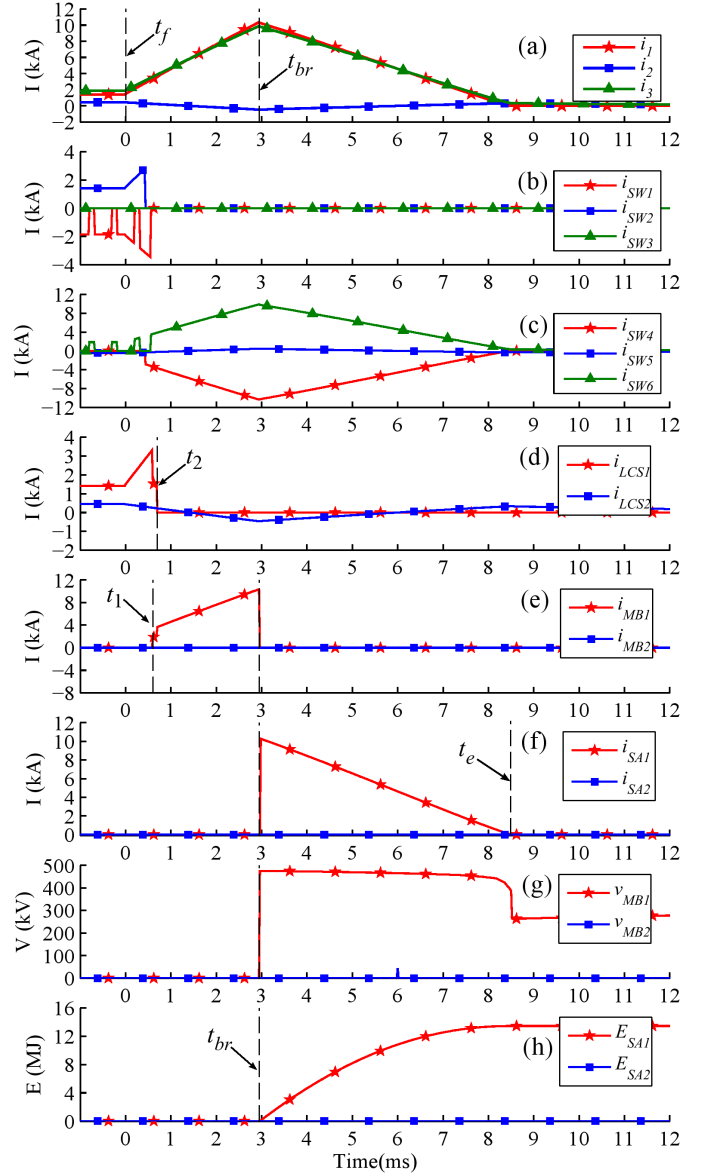
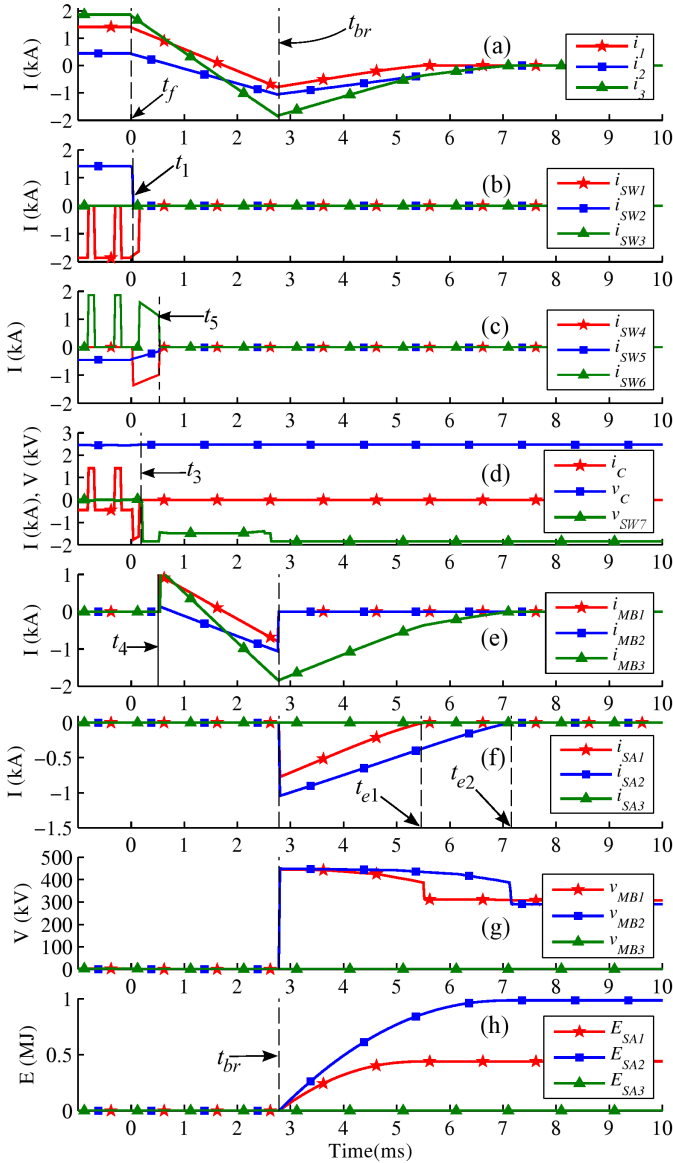
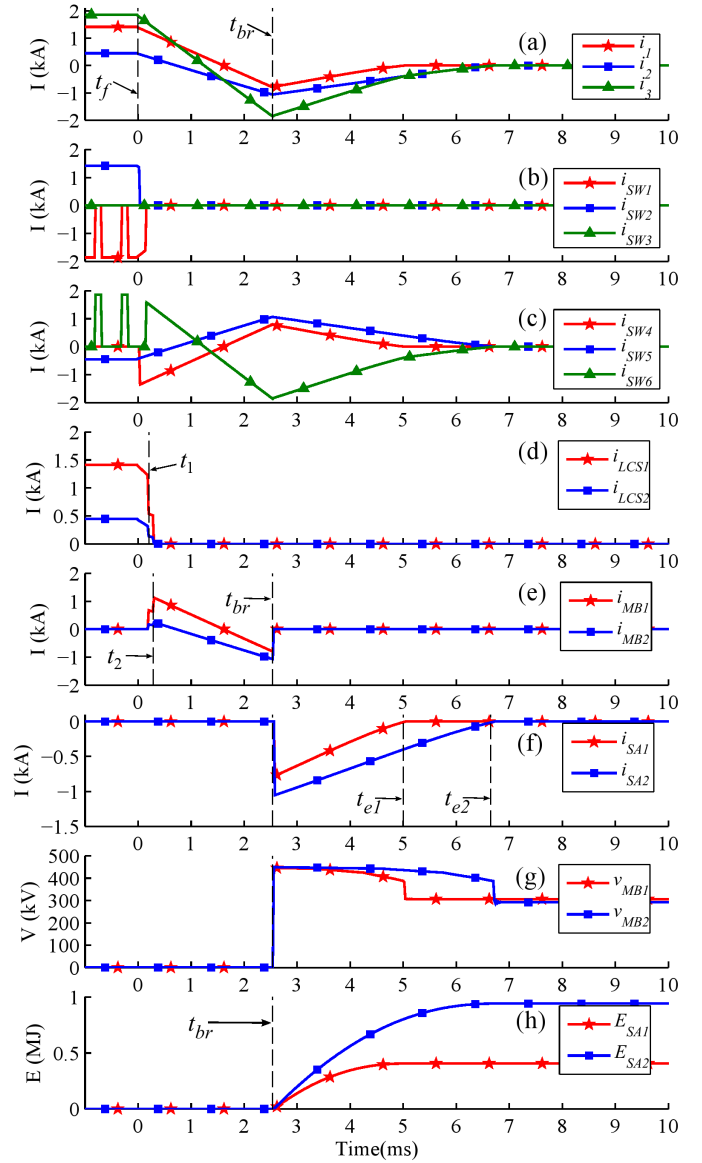


Fig. 9. Typical scheme waveforms for fault on line 12

monopole and bipole configurations. Although the comparison study is done for asymmetric monopole configuration, it is valid for other mentioned configurations. The HCBs and the CFC can be replaced by the CFCCB. The converter station HCB (CB₁₁) will not be removed. Therefore, the requirements of CB₁₁ in both cases are equal and will not be compared and included in calculations. Table IV compares different aspects of both the proposed and typical devices assuming $t_6 \approx t_{br}$.

1) *CFC*: The CFC can be bypassed during the fault either by the explained method in IV-A or using bypass valves. Considering the method from IV-A sw₄-sw₆ are required to carry the fault current till the corresponding HCBs interrupt the fault (at time t_{br}). In contrary, the same switches in the embedded CFC of CFCCB are only required to carry the fault current till the the LCS units are opened (at time t_5). In addition, sw₄-sw₆ are closed at time t_7 and share the fault current with the MB units in the line fault scenario. Comparing (30) and (13) imply that sw₄-sw₆ should be rated for higher current in the typical scheme than the proposed scheme.

Fig. 10. CFCCB waveforms for fault at bus B₁Fig. 11. Typical scheme waveforms for fault at bus B₁TABLE III
CFCCB AND HCB PARAMETERS DURING BUS FAULT

Parameters CFCCB (HCB)	CFCCB		HCB	
	Analysis	Simulation	Analysis	Simulation
$ILCS1(I_{LCS12}^{max})$ [kA]	-0.96	-0.98	-1.19	-1.24
$ILCS2(I_{LCS13}^{max})$ [kA]	-0.133	-1.151	-0.29	-0.33
$ILCS3^{max}$ (-) [kA]	-1.103	-1.13	-	-
$IMB1(I_{MB12}^{max})$ [kA]	-0.933	-0.81	-0.75	-0.78
$IMB2(I_{MB13}^{max})$ [kA]	-1.19	-1.056	-1.059	-1.06
$IMB3^{max}$ (-) [kA]	-2.12	-1.84	-	-
$E_{SA1}(E_{SA12})$ [MJ]	0.534	0.442	0.452	0.406
$E_{SA2}(E_{SA13})$ [MJ]	1.249	0.995	1.056	0.941
E_{SA3} [MJ]	0	0	-	-

A. Load commutation switches

The CFCCB uses sw₄-sw₆ of the embedded CFC as the LCSs. In fact, CFCCB saves all the IGBTs, which are required for implementing the LCS units (N_{LCS}) in the HCBs. The maximum current in LCS unit of HCB_{1j} is equal to the first maximum current in LCS_j of CFCCB for the line fault scenario. Depending on the grid topology, the second

TABLE IV
CFCCB AND HCB BASED METHODS PARAMETER COMPARISON

Parameter	HCB	CFCCB
Number of UFDs	2	2
Number of limiting ind.	2	2
Number of surge arresters	2	3
Surge arresters energy	E	$\frac{E}{2}$
Surge arresters voltage	V_r	$\frac{V_r}{2}$
Number of IGBTs in LCS	$2N_{LCS}$	0
Number of IGBTs in MB	$\frac{4V_r}{V_{CES}}$	$\frac{3V_r}{V_{CES}}$

maximum current in LCS units of CFCCB might be greater as compared to the typical HCB. During the dc bus fault condition the current in LCS₂ and LCS₃ of CFCCB are equal to the current in LCS units of HCB₁₁ to HCB₁₂. The current in LCS₁ is equal to sum of currents in LCS₂ and LCS₃ of CFCCB and therefore is higher than the currents in other units. Note that the current in LCS₁ flows through its antiparallel diodes during the bus fault. Hence, the current rating of the IGBTs are equal.

B. Main breaker units

During the line fault the current in corresponding MB units of the CFCCB and HCB are equal. Similar to the previous subsection, the antiparallel diodes of MB₁ in CFCCB may need to be able to carry higher current as compared to the other units depending on the fault identification time and the grid topology. As it is illustrated in Table IV the typical approach requires larger number of IGBTs in MB units of HCBs as compared to the proposed CFCCB. V_{CES} represents the collector-emitter blocking voltage of IGBTs in Table IV.

C. Surge arresters

1) *Rated voltage*: The rated voltage for the surge arrester of the HCB would lie in range of $1.4V_{dc} - 1.5V_{dc}$ [13], [14]. The rated voltage of surge arresters of CFCCB are also assumed to lie in the same range.

2) *Discharge current*: (14) illustrates that the maximum discharge current in the surge arresters of the CFCCB in line fault scenario is almost half of the fault current at the interruption instance. However, the maximum discharge current in the surge arrester in typical HCB is almost equal to the interrupted current. Therefore, the maximum discharge current in the surge arresters of the CFCCB is almost 50% smaller than that of the typical scheme.

3) *Energy*:

a) *Transmission line Fault*: In the typical scheme and during the line fault, only the faulty line HCB interrupts the current and its surge arrester absorbs the energy. When using the CFCCB the faulted line corresponding surge arrester does not absorb the energy and the energy absorption is shared between 2 surge arresters. Using (18) and (29) and assuming identical current at the interruption instant in both schemes the ratio of total absorbed energy in both schemes can be given as:

$$\frac{E_{SA,T}^{CFCCB}}{E_{SA,T}^{HCB}} = \frac{1}{2} \quad (33)$$

where $E_{SA,T}^{CFCCB}$ and $E_{SA,T}^{HCB}$ represents the total absorbed energy in the surge arresters of CFCCB and HCB, respectively. (33) implies that the energy rating of surge arresters in CFCCB is almost 50% smaller than that of HCB.

b) *DC bus Fault*: The CFCCB performance during the dc bus fault was found to be similar to the typical method. Therefore, equal amount of energy is absorbed in the surge arresters in both schemes.

D. Ultra-fast disconnecter

Each HCB has an ultra-fast disconnecter (UFD). As shown in Fig. 1, the CFCCB has 2 UFDs. Therefore, there is no difference in number of required UFD units for both typical and the proposed schemes.

E. Current limiting inductors

The number of current limiting inductors in both schemes are identical. Also, the inductances of current limiting inductors for the proposed and typical devices are equal.

VII. CONCLUSION

In this paper, a novel current flow controlling dc circuit breaker (CFCCB) has been proposed and analyzed. The proposed CFCCB has three ports and can connect a dc bus to two adjacent transmission lines. Each port of the CFCCB can interrupt its current, independent of the other ports and irrespective of the current direction. Furthermore, the proposed device possesses an embedded interline dual H-bridge current flow controller (CFC), which can regulate the current in one of the adjacent transmission lines.

While the proposed device shows similar behavior compared to the typical scheme from the system level view, it can reduce the requirements of different elements of system. The analysis and simulations imply that the proposed CFCCB requires fewer IGBTs compared to the typical approach. For a dc bus with two adjacent transmission lines, CFCCB needs at least 25% fewer IGBTs as compared to the typical scheme. Moreover, the proposed device requires smaller size surge arresters due to the less energy absorption in its surge arresters. In addition, the maximum current discharge in the surge arresters in the proposed method is smaller than that of typical method. The results from this study confirm that the energy and discharge current ratings of the surge arresters can be reduced by almost 50%. Considering the improvements by applying the proposed device, its implementation cost is expected to be remarkably lower than the cost of typical scheme. The future work will concern with the cost-benefit and reliability studies of the proposed device.

APPENDIX

TABLE V
THREE-TERMINAL TEST GRID AND CFC PARAMETERS

Transmission Lines Parameters			
Lumped T-model Parameters	R [Ω /km]	L [mH/km]	C [μ F/km]
	0.01105	3.245	0.382
Length [km]	Line 12	Line 13	Line 23
	200	300	200
VSC Parameters			
Bus	1	2	3
Capacitance [μ F]	1000	1000	1000
Power [MW]	600	-	-
V_i^* [kV]	-	300	300
Droop Constant k_i [A/V]	-	0.05	0.5
Voltage [kV]	320	320	320
Filter reactor [mH]	10	10	10
CFC Parameters			
Nominal Voltage [kV]	4		
Capacitor [mF]	10		
Switching Frequency [kHz]	2		
PI	$0.012 + \frac{0.398}{s}$		
Compensator	$\frac{0.3421s^2 + 1.2978s + 21.5213}{s^2 + 120.5323s + 3207.9071}$		
Filter	$\frac{1}{0.08s + 1}$		

REFERENCES

- [1] D. V. Hertem, O. Gomis-Bellmunt, and J. Liang, *HVDC GRIDS For Offshore and Supergrid of the Future*. John Wiley & Sons, Inc., 2016.
- [2] B. C. N. Chaudhuri, R. Majumder, and A. Yazdani, *Multi-terminal DirectCurrent Grids: Modeling, Analysis, and Control*. John Wiley & Sons, Inc., 2014.
- [3] A. Mokhberdorani, N. Silva, H. Leite, and A. Carvalho, "Unidirectional Protection Strategy for Multi-terminal HVDC Grids," *Transactions on Environment and Electrical Engineering*, vol. 1, no. 4, pp. 58–65, 2016.

- [4] A. Mokhberdorán and A. Ajami, "Symmetric and asymmetric design and implementation of new cascaded multilevel inverter topology," *IEEE Transactions on Power Electronics*, vol. 29, no. 12, pp. 6712–6724, Dec 2014.
- [5] C. M. Franck, "Hvdc circuit breakers: A review identifying future research needs," *IEEE Transactions on Power Delivery*, vol. 26, no. 2, pp. 998–1007, April 2011.
- [6] A. Mokhberdorán, A. Carvalho, H. Leite, and N. Silva, "A review on hvdc circuit breakers," in *Renewable Power Generation Conference (RPG 2014)*, 3rd, Sept 2014, pp. 1–6.
- [7] J. Sau-Bassols, E. Prieto-Araujo, and O. Gomis-Bellmunt, "Modelling and control of an interline current flow controller for meshed hvdc grids," *IEEE Transactions on Power Delivery*, vol. PP, no. 99, pp. 1–1, 2016.
- [8] K. A. Corzine, "A new-coupled-inductor circuit breaker for dc applications," *IEEE Transactions on Power Electronics*, vol. 32, no. 2, pp. 1411–1418, Feb 2017.
- [9] D. Keshavarzi, T. Ghanbari, and E. Farjah, "A z-source based bidirectional dc circuit breaker with fault current limitation and interruption capabilities," *IEEE Transactions on Power Electronics*, vol. PP, no. 99, pp. 1–1, 2016.
- [10] A. Maqsood, A. Overstreet, and K. A. Corzine, "Modified z-source dc circuit breaker topologies," *IEEE Transactions on Power Electronics*, vol. 31, no. 10, pp. 7394–7403, Oct 2016.
- [11] A. Mokhberdorán, A. Carvalho, N. Silva, H. Leite, and A. Carrapatoso, "A new topology of fast solid-state hvdc circuit breaker for offshore wind integration applications," in *Power Electronics and Applications (EPE'15 ECCE-Europe)*, 2015 17th European Conference on, Sept 2015, pp. 1–10.
- [12] J. Hafner and B. Jacobson, "Proactive hybrid hvdc breakers - a key innovation for reliable hvdc grids," in *Electric power system of the future - Integrating supergrids and microgrids international symposium*, 2011.
- [13] A. Hassanpoor, J. Hafner, and B. Jacobson, "Technical assessment of load commutation switch in hybrid hvdc breaker," *IEEE Transactions on Power Electronics*, vol. 30, no. 10, pp. 5393–5400, Oct 2015.
- [14] G. Liu, F. Xu, Z. Xu, Z. Zhang, and G. Tang, "Assembly hvdc breaker for hvdc grids with modular multilevel converters," *IEEE Transactions on Power Electronics*, vol. 32, no. 2, pp. 931–941, Feb 2017.
- [15] A. Mokhberdorán, A. Carvalho, N. Silva, H. Leite, and A. Carrapatoso, "Application study of superconducting fault current limiters in meshed hvdc grids protected by fast protection relays," *Electric Power Systems Research*, vol. 143, pp. 292 – 302, 2017.
- [16] C. D. Barker and R. S. Whitehouse, "A current flow controller for use in hvdc grids," in *AC and DC Power Transmission (ACDC 2012)*, 10th IET International Conference on, Dec 2012, pp. 1–5.
- [17] D. Jovicic, M. Hajian, H. Zhang, and G. Asplund, "Power flow control in dc transmission grids using mechanical and semiconductor based dc/dc devices," in *AC and DC Power Transmission (ACDC 2012)*, 10th IET International Conference on, Dec 2012, pp. 1–6.
- [18] S. Balasubramaniam, J. Liang, and C. E. UgaldeLoo, "Control, dynamics and operation of a dual h-bridge current flow controller," in *2015 IEEE Energy Conversion Congress and Exposition (ECCE)*, Sept 2015, pp. 2386–2393.
- [19] V. Hofmann, A. Schn, and M. M. Bakran, "A modular and scalable hvdc current flow controller," in *Power Electronics and Applications (EPE'15 ECCE-Europe)*, 2015 17th European Conference on, Sept 2015, pp. 1–9.
- [20] M. Ranjram and P. W. Lehn, "A multiport power-flow controller for dc transmission grids," *IEEE Transactions on Power Delivery*, vol. 31, no. 1, pp. 389–396, Feb 2016.
- [21] W. Chen, X. Zhu, L. Yao, G. Ning, Y. Li, Z. Wang, W. Gu, and X. Qu, "A novel interline dc power-flow controller (idcpfc) for meshed hvdc grids," *IEEE Transactions on Power Delivery*, vol. 31, no. 4, pp. 1719–1727, Aug 2016.
- [22] W. Leterme and D. V. Hertem, "Classification of fault clearing strategies for hvdc grids," in *presented at the 2015 Lund Symposium, Cigre, Lund*, 2015.
- [23] D. Jovicic and K. Ahmed, *High Voltage Direct Current Transmission: Converters, System and DC Grid*. John Wiley & Sons, Inc., 2015.
- [24] W. Lin, D. Jovicic, S. Nguefeu, and H. Saad, "Modelling of high-power hybrid dc circuit breaker for grid-level studies," *IET Power Electronics*, vol. 9, no. 2, pp. 237–246, 2016.



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